

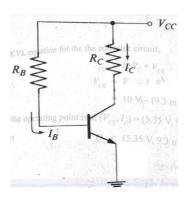


Internal Assesment Test - I

Sub:	ANALOG ELECTRONIC CIRCUITS Cod							e:	18EE34		
Date	: 09/09/2019	Duration:	90 mins	Max Marks:	50	Sem:	3rd	Bran	ch:	EEI	Ξ
Answer Any FIVE FULL Questions											
								Marks	arks OBE CO RBT		
1	For the fixed-bias circu	it, $R_B=50K\Omega$	e, Rc=5009	$0, V_{CC} = 10V.$					10	CO1	L3
		eat circuit dia									
	b) Find the coordinates of the operating point.										
	c) Draw the dc load line and locate the operating point on the dc load line.										
	Assume silicon transistor with $\beta = 50$ and $V_{BE} = 0.7V$. The circuit shown below uses silicon transistor. Calculate								10	CO1	1.2
2									10	CO1	L3
	a) β (b) V_{cc}		- 75	;)							
	, x 6		V _{CC}	a							
			1								
		Va.	1	> 2.7 kΩ							
		$R_B \lessapprox \downarrow 20 $		2.7 KS2	36						
		> 20 F		+	V_C						
	=	N - 1		7.3 V							
	<i>₩</i> .		45.	51,7	2.1 V						
		(100) ¹²	1		,						
	2200		680 Ω	\geq							
			· · · · · · · · ·								
3	Derive the expression for	or stability fa	ctor $S(I_{co})$	and $S(V_{BE})$	for Emit	ter Bias			10	CO1	L3
4	configuration.	C (17	1.1	1 . 1 .	2506	1. 400	000	.1	10	CO1	L3
4	Determine the stability factor $S(V_{BE})$ and the change in I_C from $25^{\circ}C$ to $100^{\circ}C$ for the							r the	10	COI	LS
	transistor with $V_{BE}(25^{\circ}C) = 0.65V$ and $V_{BE}(100^{\circ}C) = 0.48V$ for the following bias							bias			
	arrangements. a) Fixed bias with $R_B = 270k\Omega$ and $\beta = 120$										
	b) Voltage divider bias				= 1 <i>kO</i> a	nd R =	120				
	Derive the expression for							ation	10	CO1	L3
							migui	ation			
6	For the transistor invert			_	_				10	CO1	L3
	$I_{C(sat)} = 12mA, \beta =$	= 200 and V_c	E(sat) =	0 V. Also draw	the outp	ut voltag	ge				
	Waveform.	1/		12 V							
			17.	,							
	/			$\stackrel{\downarrow}{\geqslant}_{R_C}$							
	12	v		\ \{\bar{\}}^{\infty}							
	0	v		~ ,	∘ V _C						
	_		R_B	1	¥ * 0						
		V	, °								
				1							
7	What is Clamping circu	its? Draw the	e circuit ar	nd output wavef	orms for	negativ	e clan	nper	10	CO1	L1
	and positive clamper.			•				^			

1.

a) Circuit diagram:



b)

$$I_B = \frac{V_{CC} - V_{BE}}{R_B} = \frac{10 \text{ V} - 0.7 \text{ V}}{50 \text{ k}\Omega} = 0.186 \text{ mA}$$

$$I_C = \beta I_B = 50 \times 0.186 \text{ mA} = 9.3 \text{ mA}$$

Writing KVL equation for the the collector circuit,

$$V_{CC} = I_C R_C + V_{CE}$$

$$V_{CE} = V_{CC} - I_C R_C$$

$$= 10 \text{ V} - (9.3 \text{ mA} \times 0.5 \text{ k}\Omega) = 5.35 \text{ V}$$
ting point is at $(V_{CE}, I_C) = (5.35 \text{ V}, 9.3 \text{ mA})$

Therefore the operating point is at $(V_{CE}, I_C) = (5.35 \text{ V}, 9.3 \text{ mA})$

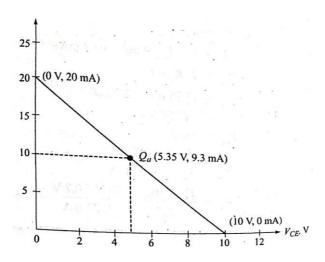
Let
$$Q_a = (5.35 \text{ V}, 9.3 \text{ mA})$$

The current axis intercept of the load line is

$$\frac{V_{\rm CC}}{R_{\rm C}} = \frac{10\,\mathrm{V}}{0.5\,\mathrm{k}\Omega} = 20\,\mathrm{mA}$$

and the voltage axis intercept is $V_{CC} = 10$ V. The load line is plotted by joining ((and (10 V, 0 mA) as shown below:

c)



(a) β Given the voltage across the emitter resistance, we can obtain the emitter current at

$$I_E = \frac{V_E}{R_E} = \frac{2.1 \text{ V}}{680 \Omega} = 3.09 \text{ mA}$$

The emitter current is approximately equal to collector current.

$$I_{C} = 3.09 \text{ mA}$$
Now
$$\beta = \frac{I_{C}}{I_{B}} = \frac{3.09 \text{ mA}}{20 \text{ \mu A}} = 154.5$$

(b) V_{cc} From the KVL equation for the collector emitter circuit, we have

$$V_{CC} = I_C R_C + V_{CE} + V_E$$

= (3.09 mA × 2.7 k\Omega) + 7.3 V + 2.1 V = 17.74

Writing the KVL equation for the base-emitter circuit

$$R_{B} = \frac{V_{CC} - V_{BE} - V_{E}}{I_{B}}$$

$$= \frac{17.74 \text{ V} - 0.7 \text{ V} - 2.1 \text{ V}}{20 \text{ } \mu\text{A}} = 747 \text{ } k\Omega$$

$$V_C = V_{CE} + V_E = 7.3 \text{ V} + 2.1 \text{ V} = 9.4 \text{ V}$$

e)
$$I_{C \text{ (sat)}} = \frac{V_{CC}}{R_C + R_E} = \frac{17.74 \text{ V}}{2.7 \text{ k}\Omega + 680 \Omega} = 5.248 \text{ mA}$$

3. Stability factor $S(I_{co})$

d)

$$S(I_{CO}) = \frac{1+\beta}{1-\beta} \frac{\partial I_B}{\partial I_C}$$

$$V_{CC} = I_B R_B + V_{BE} + I_E R_E$$
But
$$I_E = I_B + I_C$$

$$V_{CC} = I_B R_B + V_{BE} + I_B R_E + I_C R_E$$

Differentiating

$$0 = \frac{\partial I_B}{\partial I_C} R_B + \frac{\partial I_B}{\partial I_C} R_E + R_E = \frac{\partial I_B}{\partial I_C} (R_B + R_E) + R_E$$

$$\frac{\partial I_B}{\partial I_C} = -\frac{R_E}{R_B + R_E}$$

$$S(I_{CO}) = \frac{1+\beta}{1-\beta \left(\frac{-R_E}{R_B + R_E}\right)} = \frac{1+\beta}{1+\beta \left(\frac{R_E}{R_B + R_E}\right)}$$

$$= \frac{(\beta+1)(R_E + R_B)}{R_B + R_E + \beta R_E} = \frac{(\beta+1)(R_E + R_B)}{(\beta+1)R_E + R_B}$$

$$S(I_{CO}) = \frac{(\beta+1)\left(1+\frac{R_B}{R_E}\right)}{(\beta+1)+\frac{R_B}{R_E}}$$

Stability factor S(V_{BE})

$$S(V_{BE}) = \frac{\partial I_C}{\partial V_{BE}}$$

$$V_{CC} = I_B R_B + V_{BE} + I_B R_E + I_C R_E$$

$$V_{CC} = \frac{I_C}{\beta} R_B + V_{BE} + \frac{I_C}{\beta} R_E + I_C R_E$$

$$= \left[\frac{R_B}{\beta} + \frac{R_E}{\beta} + R_E \right] I_C + V_{BE}$$

$$= \left[\frac{R_B + R_E + \beta R_E}{\beta} \right] I_C + V_{BE}$$

$$V_{CC} = \left[\frac{R_B + (1+\beta)R_E}{\beta} \right] I_C + V_{BE}$$

$$0 = \frac{R_B + (1+\beta)R_E}{\beta} + \frac{\partial V_{BE}}{\partial I_C}$$

$$S(V_{BE}) = \frac{\partial I_C}{\partial V_{BE}} = \frac{-\beta}{R_B + (1+\beta)R_E}$$

4.

$$\Delta V_{BE} = V_{BE} (100^{\circ} \text{ C}) - V_{BE} (25^{\circ} \text{ C})$$

= 0.48 V - 0.65 V = -0.17 V

a)

Fixed Bias
$$k_1 = 270 \text{ k}\Omega \quad \beta = 120$$

$$S(V_{BE}) = \frac{-\beta}{R_B} = -\frac{120}{270 \text{ k}\Omega} = -0.44 \times 10^{-3}$$
$$\Delta I_C = S(V_{BE}) \Delta V_{BE}$$
$$= (-0.44 \times 10^{-3}) (-0.17) = 74.8 \text{ } \mu\text{A}$$

b)

Voltage Divider Bias
$$R_{1} = 39 \text{ k}\Omega \quad R_{2} = 10 \text{ k}\Omega \quad R_{E} = 1\text{k}\Omega \quad \beta = 120$$

$$S(V_{BE}) = \frac{-\beta}{R_{Th} + (1+\beta)R_{E}}$$

$$R_{th} = R_{1} \parallel R_{2} = 39 \text{ k}\Omega \parallel 10 \text{ k}\Omega = 7.95 \text{ k}\Omega$$

$$S(V_{BE}) = \frac{-120}{7.95 \text{ k}\Omega + (121)(1 \text{ k}\Omega)} = -0.93 \times 10^{-3}$$

$$\Delta I_{C} = S(V_{BE}) \Delta V_{BE} = (-0.93 \times 10^{-3}) (-0.17) = 158.1 \text{ p}$$

5. Stability factor $S(I_{co})$

$$S(I_{co}) \equiv \frac{\partial I_{ci}}{\partial I_{co}}$$

$$I_{B} = \frac{V_{cc} - V_{BE}}{R_{B}}$$

But

$$V_{CC} \gg V_{BE}$$

$$I_{B} \simeq \frac{V_{CC}}{R_{B}} \quad \text{which is constant}$$

Differentiating with respect to I_C ,

$$\frac{\partial I_B}{\partial I_C} = 0$$

$$S(I_{CO}) = 1 + \beta$$

Stability factor $S(V_{BE})$

$$S(V_{BE}) \equiv \frac{\partial I_{C}}{\partial V_{BE}}$$

$$V_{CC} = I_{B}R_{B} + V_{BE}$$

$$V_{BE} = V_{CC} - I_{B}R_{B}$$

$$I_{B} \simeq \frac{I_{C}}{\beta}$$

$$V_{BE} = V_{CC} - \frac{I_{C}}{\beta} R_{B}$$

$$1 = 0 - \frac{R_{B}}{\beta} \frac{\partial I_{C}}{\partial V_{BE}}$$

$$S(V_{BE}) = -\frac{\beta}{R_B}$$

6.

Calculation of R_C

$$I_{C\,(\text{sat})} = \frac{V_{CC} - V_{CE\,(\text{sat})}}{R_C}$$

From the output waveform, $V_{CE(sat)} = 0 \text{ V}$

$$R_C = \frac{V_{CC}}{I_{C(\text{sat})}} = \frac{12 \text{ V}}{12 \text{ mA}} = 1 \text{ k}\Omega$$

Calculation of R.

$$I_{B(\text{max})} = \frac{I_{C(\text{sat})}}{\beta_{\text{dc}}} = \frac{12 \text{ mA}}{200} = 60 \text{ }\mu\text{A}$$
Let
$$I_{B} = 150 \% \text{ of } I_{B(\text{max})} \quad \text{[To ensure saturation]}$$

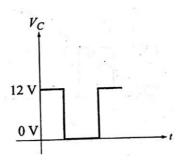
$$= (1.5) (60 \text{ }\mu\text{A}) = 90 \text{ }\mu\text{A}$$

$$R_{B} = \frac{V_{i} - V_{BE}}{I_{B}}$$

$$= \frac{12 \text{ V} - 0.7 \text{ V}}{90 \text{ }\mu\text{A}}$$

$$= 125.55 \text{ k}\Omega$$

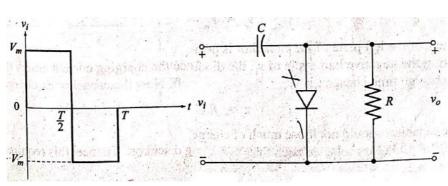
Output voltage waveform

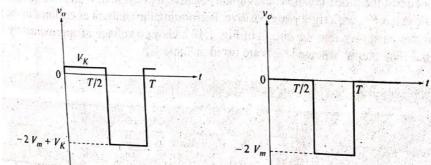


7.

Clamping circuits are used toad dc level to the input signal. They are also called dc restorers or dc inserters. It uses diode, resistors and capacitors.

Negative clamper





Positive clamper

