

Solutions of Internal Assessment Test - I

Sub:	DIGITAL SYSTEM DESIGN		Code:	18EE35
Sem:	3rd	Branch:	EEE	

1. Distinguish prime implicants and essential prime implicants. Determine prime implicants and essential prime implicants of $Y=f(a,b,c,d)=\pi M(0,1,4,5,8,9,11)+d(2,10)$ and obtain the final expression.

Solution:

→ Prime Implicants	Essential Prime Implicants																
<p>① In K-map, after grouping the cells, the sum terms having the minterm that appears are called prime implicant groups. And the minterms are known as prime implicant.</p> <p>② All prime implicants are not essential prime implicants.</p> <p>③ Example: AB</p> <table border="1" style="display: inline-table; vertical-align: middle;"> <tr><td>0</td><td>1</td><td>0</td><td>0</td></tr> <tr><td>1</td><td>0</td><td>0</td><td>0</td></tr> <tr><td>0</td><td>0</td><td>0</td><td>1</td></tr> <tr><td>1</td><td>1</td><td>0</td><td>1</td></tr> </table> <p>$f(A,B,C,D)$ $= \overline{A}\overline{C}\overline{D} + \overline{A}\overline{B}\overline{D} + \overline{B}\overline{C}$</p> <p>Cell 0, 1, 4, 5, 8, 9, 10 are all from prime implicant group.</p>	0	1	0	0	1	0	0	0	0	0	0	1	1	1	0	1	<p>① In K-map, after grouping the cell, it is observed that some cells may appear in only one prime implicant groups. These are known as essential prime implicants.</p> <p>② All essential prime implicants are the prime implicants.</p> <p>③ Example: from K-map cells, 1, 4, 9, 10 are appeared only in one prime implicants group.</p>
0	1	0	0														
1	0	0	0														
0	0	0	1														
1	1	0	1														

Solution : Prime implicants :

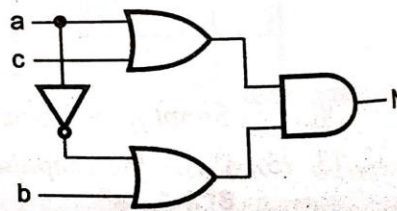
		cd		
ab	c+d	c+d	c̄+d̄	c̄+d
a+b	0	0		x
a+b̄	0	0		
ā+b				
ā+b̄	0	0	0	x

$$N = (a + c)(\bar{a} + b)(b + c)$$

Essential prime implicants :

$$N = (a + c)(\bar{a} + b)$$

Implementation :



2. Find a minimum SOP solution using Quine-McCluskey method
 $F(a,b,c,d) = \sum m(2,3,4,5,13,15) + d(8,9,10,11)$.

Solution:

Step 1 : List all minterms in binary form.

Minterms	Binary representation			
m_2	0	0	1	0
m_3	0	0	1	1
m_4	0	1	0	0
m_5	0	1	0	1
m_{13}	1	1	0	1
m_{15}	1	1	1	1
dm_8	1	0	0	0
dm_9	1	0	0	1
dm_{10}	1	0	1	0
dm_{11}	1	0	1	1

Step 2 : Arrange the minterms according to number of 1's.

Minterms	Binary representation			
m_2	0	0	1	0
m_4	0	1	0	0
m_8	1	0	0	0
m_3	0	0	1	1
m_5	0	1	0	1
m_9	1	0	0	1
m_{10}	1	0	1	0
m_{11}	1	0	1	1
m_{13}	1	1	0	1
m_{15}	1	1	1	1

Step 3 :

Minterm	Binary representation			
2, 3 ✓	0	0	1	-
2, 10 ✓	-	0	1	0
4, 5	0	1	0	-
8, 10 ✓	1	0	-	0
8, 9 ✓	1	0	0	-
3, 11 ✓	-	0	1	1
5, 13	-	1	0	1
9, 13 ✓	1	-	0	1
9, 11 ✓	1	0	-	1
10, 11 ✓	1	0	1	-
13, 15 ✓	1	1	-	1
11, 15 ✓	1	-	1	1

Step 4 :

Minterm	Binary representation			
2, 3, 10, 11	-	0	1	-
8, 9, 10, 11	1	0	-	-
9, 11, 13, 15	1	-	-	1

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Step 5 :

Prime implicants				Binary representation			
\bar{a}	b	\bar{c}	4, 5	0	1	0	-
b	\bar{c}	d	5, 13	-	1	0	1
\bar{b}	c		2, 3, 10, 11	-	0	1	-
a	\bar{b}		8, 9, 10, 11	1	0	-	-
a	d		9, 11, 13, 15	1	-	-	1

Step 6

Prime Implicants	m ₂	m ₃	m ₄	m ₅	m ₁₃	m ₁₅	dm ₈	dm ₉	dm ₁₀	dm ₁₁
4, 5 $\bar{a} b \bar{c}$			⊙	•						
5, 13 $b \bar{c} d$				•	•					
2, 3, 10, 11 $\bar{b} c$	⊙	⊙							•	•
8, 9, 10, 11 $a \bar{b}$							•	•	•	•
9, 11, 13, 15 ad					•	⊙		•		•

$$\therefore Y = \bar{b}c + \bar{a}b\bar{c} + ad$$

3. Place the following equations into proper canonical forms:

(a) $P=f(a,b,c) = ab'+ab'+bc$

(b) $T=f(a,b,c) = (a+b')(b'+c)$

Solution:

$$\begin{aligned}
 (a) \quad P = f(a, b, c) &= ab' + ab' + bc \\
 &= ab'(c+c') + ab'(c+c') + bc(a+a') \\
 &= ab'c + ab'c' + ab'c + ab'c' + abc + a'bc \\
 \therefore f(a, b, c) &= ab'c + ab'c' + abc + a'bc
 \end{aligned}$$

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↑
Canonical Form.

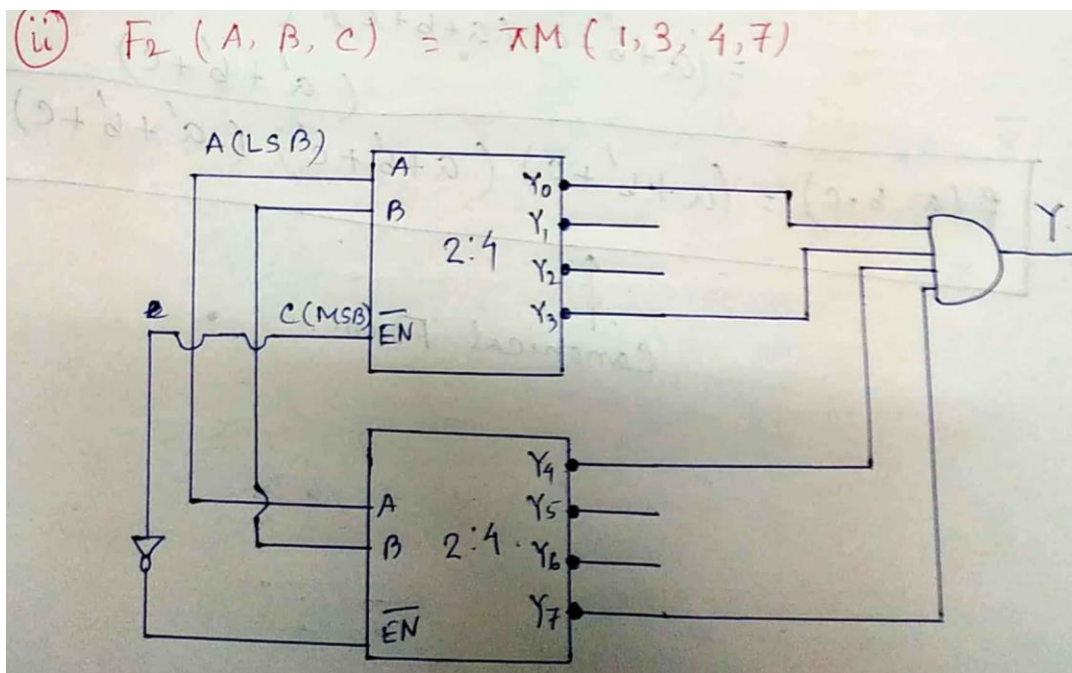
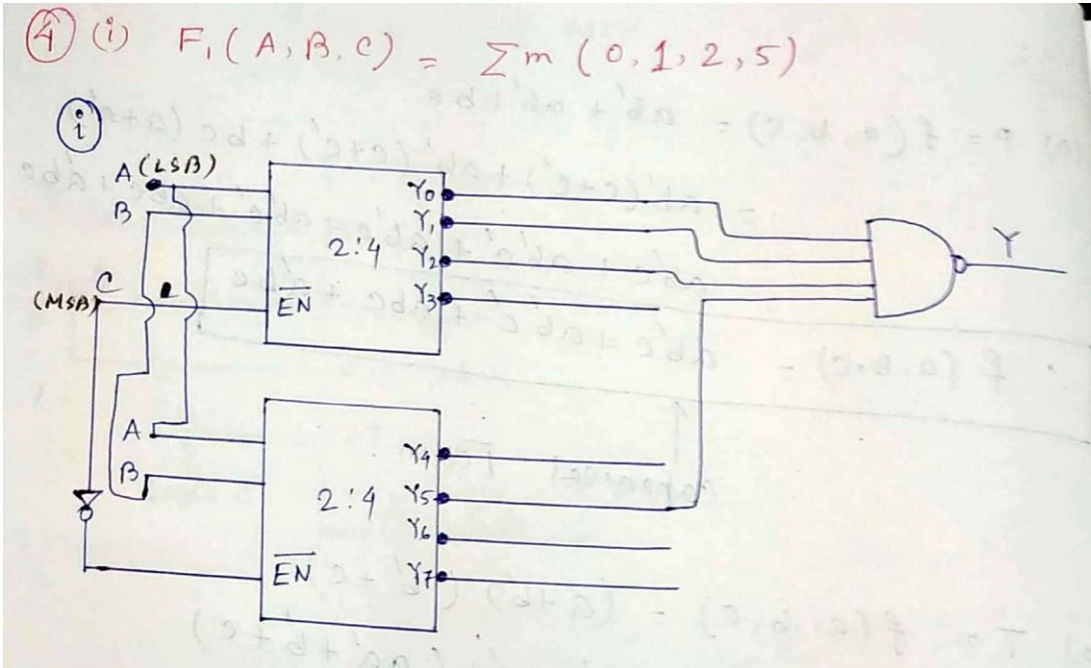
$$\begin{aligned}
 (b) \quad T = f(a, b, c) &= (a+b')(b'+c) \\
 &= (a+b'+cc')(a'+b'+c) \\
 &= (a+b'+c)(a+b'+c')(a'+b'+c) \\
 f(a, b, c) &= (a+b'+c)(a+b'+c')(a'+b'+c)
 \end{aligned}$$

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↑
Canonical Form.

4. Implement the function using active low output dual 2:4 decoder line decoder IC 74139
 i) $F_1(A,B,C) = \sum m(0,1,2,5)$ ii) $F_2(A,B,C) = \pi M(1,3,4,7)$

Solution:



5. What is the problem associated with the parallel adder? Explain the method of correcting it, with suitable circuit equations.

Solution:

The parallel adder is ripple carry adder in which the carry output of each full-adder stage is connected to the carry input of the next higher-order stage. Therefore, the sum and carry outputs of any stage cannot be produced until the input carry occurs; this leads to a time delay in the addition process. This delay is known as **carry propagation delay**.

One method of speeding up this process by eliminating inter stage carry delay is called **look ahead-carry addition**. This method utilizes logic gates to look at the lower-order bits of the augend and addend to see if a higher-order carry is to be generated. It uses two functions : carry generate and carry propagate.

Consider the circuit of the full-adder shown in Fig. 2.11.1. Here, we define two functions : carry generate and carry propagate.

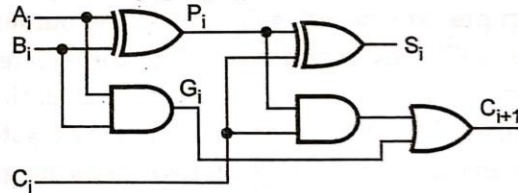


Fig. 2.11.1 Full-adder circuit

$$P_i = A_i \oplus B_i$$

$$G_i = A_i B_i$$

The output sum and carry can be expressed as

$$S_i = P_i \oplus C_i$$

$$C_{i+1} = G_i + P_i C_i$$

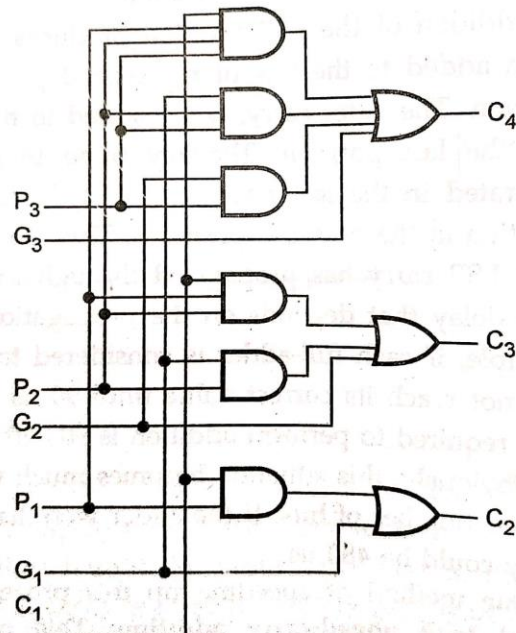


Fig. 2.11.2 Logic diagram of a look ahead carry generator

G_i is called a carry generate and it produces on carry when both A_i and B_i are one, regardless of the input carry. P_i is called a carry propagate because it is term associated with the propagation of the carry from C_i to C_{i+1} .

Now the Boolean function for the carry output of each stage can be written as follows.

$$C_2 = G_1 + P_1 C_1$$

$$C_3 = G_2 + P_2 C_2 = G_2 + P_2 (G_1 + P_1 C_1) = G_2 + P_2 G_1 + P_2 P_1 C_1$$

$$C_4 = G_3 + P_3 C_3 = G_3 + P_3 (G_2 + P_2 G_1 + P_2 P_1 C_1)$$

$$= G_3 + P_3 G_2 + P_3 P_2 G_1 + P_3 P_2 P_1 C_1$$



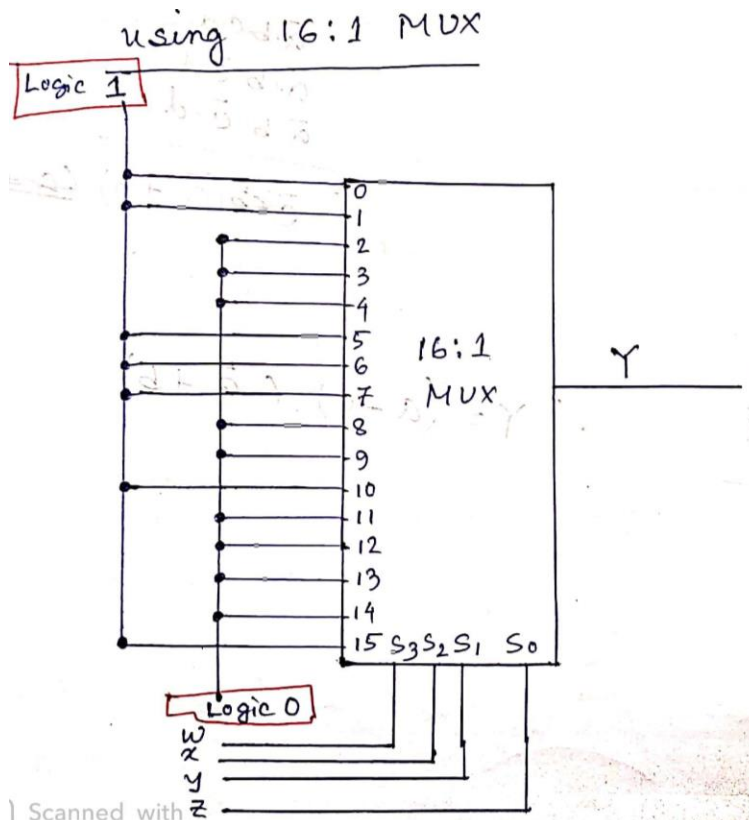
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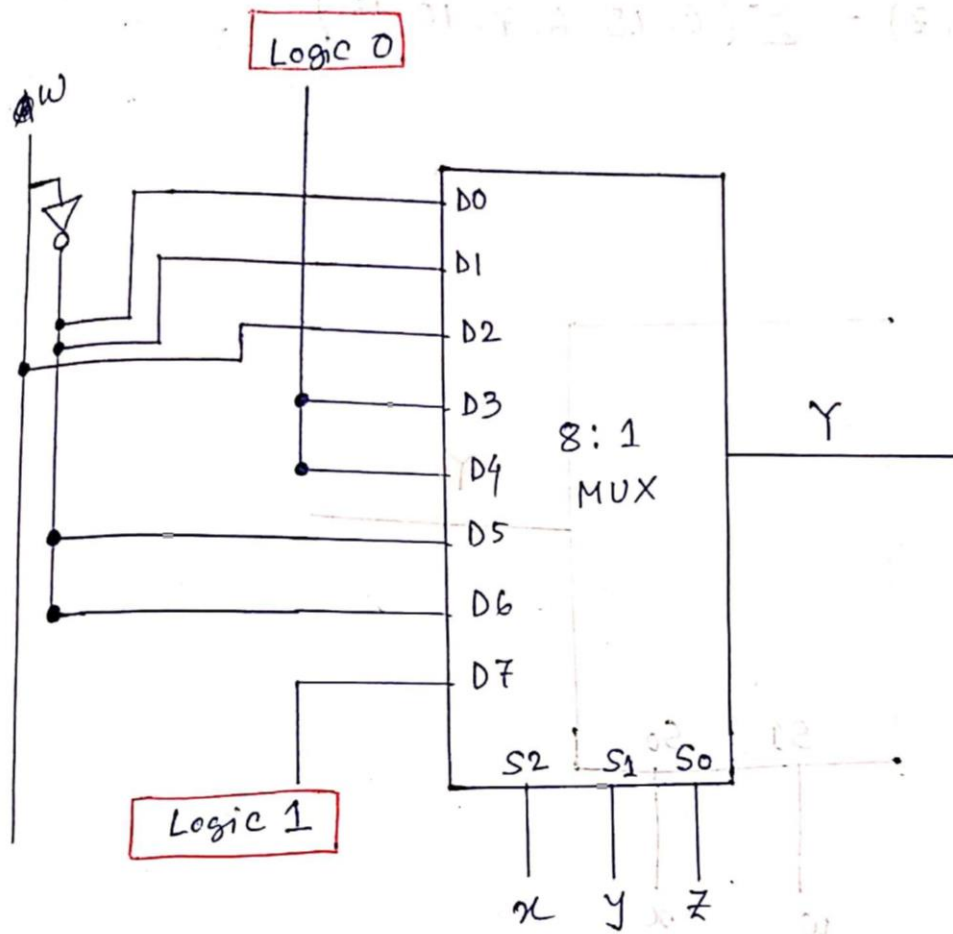
From the above Boolean function it can be seen that C_4 does not have to wait for C_3 and C_2 to propagate; in fact C_4 is propagated at the same time as C_2 and C_3 .

6. Realize the following Boolean function $Y=f(w,x,y,z)=\sum(0,1,5,6,7,10,15)$ using :

- i) 16 to 1 MUX ii) 8:1 MUX iii) 4:1 MUX.

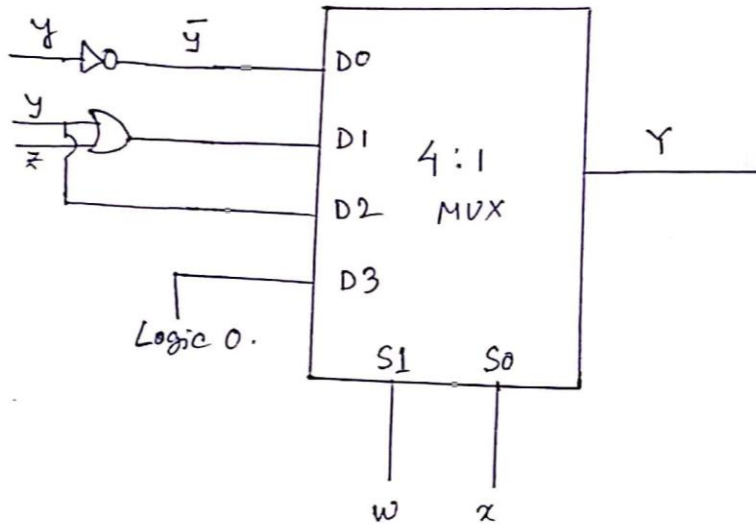
Solution:





	D0	D1	D2	D3	D4	D5	D6	D7
\bar{w}	0	1	2	3	4	5	6	7
w	8	9	10	11	12	13	14	15
	\bar{w}	\bar{w}	w	0	0	\bar{w}	\bar{w}	1

Using 4:1 MUX



	$\bar{y}\bar{z}$	$\bar{y}z$	$y\bar{z}$	yz	
D0	0	1	2	3	$\bar{y}\bar{z} + \bar{y}z = \bar{y}$
D1	4	5	6	7	$\bar{y}z + y\bar{z} + yz$
D2	8	9	10	11	$y\bar{z} + yz$
D3	12	13	14	15	0

$$\begin{aligned} \therefore D0 &= \bar{y} \\ D1 &= \bar{y}\bar{z} + y\bar{z} + yz \\ &= \bar{y}\bar{z} + y \\ &= y + \bar{z} \\ D2 &= y\bar{z} + yz \\ D3 &= 0 = y \end{aligned}$$