


CMR INSTITUTE OF TECHNOLOGY		USN											
Internal Assessment Test – 2													
<b>Sub:</b>	Microwave and Antennas							<b>Code:</b>	18EC63				
<b>Date:</b>	09/06/2022	<b>Duration:</b>	90 mins	<b>Max Marks:</b>	50	<b>Sem:</b>	6	<b>Branch:</b>	ECE				
<b>Answer Any FIVE FULL Questions</b>													
<b>Questions</b>								<b>Marks</b>	<b>CO</b>	<b>RBT</b>			
1	(a) Prove that it is not possible to construct to perfectly matched lossless reciprocal three port junction.  (b) Derive S-matrix of E-plane Tee junction							[10]	CO2	L4			
2	What is S-parameter? State and prove unitary properties of S-matrix.							[10]	CO2	L2			
3	(a) A 20 mW signal is fed into the collinear arm of a lossless H-Plane Tee. Calculate the Power delivered through each port when others ports are terminated with matched load.  (b) Define the following losses in a microwave network in terms of S – parameters: (i) Insertion loss, (ii) Transmission loss, (iii) Reflection loss, (iv) Return loss.							[10]	CO2	L3			
4	What is a magic tee? Derive the S-matrix of magic tee. Mention its applications.							[10]	CO2	L3			
5	With the help of a neat diagram, explain the working of a precision phase shifter.  <b>OR</b>  With the help of a neat diagram, explain the working of an attenuator.							[10]	CO2	L2			
6.	Drive the relation between [S] and [Z] Matrix							[10]	CO3	L3			
7.	Drive the relation between [S] and ABCD Parameter Matrix							[10]	CO3	L3			
8.	Two transmission line of characteristic impedance $Z$ and $Z_2$ are joined at plane $pp'$ . Express S-parameter in terms of impedances when each line is matched terminated.							[10]	CO3	L3			

Q1.

(a) Consider a  $S$ -matrix,

$$S = \begin{pmatrix} S_{11} & S_{12} & S_{13} \\ S_{21} & S_{22} & S_{23} \\ S_{31} & S_{32} & S_{33} \end{pmatrix}$$

Perfectly  
matched  $\rightarrow$  
$$\begin{pmatrix} 0 & S_{12} & S_{13} \\ S_{21} & 0 & S_{23} \\ S_{31} & S_{32} & 0 \end{pmatrix} \quad [ \because \text{diagonal elements are zero} ]$$

Reciprocal  
Condition  $\rightarrow$  
$$\begin{pmatrix} 0 & S_{12} & S_{13} \\ S_{12} & 0 & S_{23} \\ S_{13} & S_{23} & 0 \end{pmatrix} \quad [ \because S_{ij} = S_{ji} ]$$

$$\therefore S_{12} = S_{21} ; S_{13} = S_{31} ; S_{23} = S_{23}$$

$$\sum_{k=1}^N S_{ki} S_{kj}^* = 0$$

$\rightarrow$  Consider any two columns

$$0 \times S_{12}^* + S_{12} \times 0 + S_{13} \times S_{23}^* = 0 \quad \text{--- (1)}$$

$$0 \times S_{13}^* + S_{12} \times S_{23}^* + S_{13} \times 0 = 0 \quad \text{--- (2)}$$

$$S_{12} \times S_{13}^* + 0 \times S_{23}^* + S_{23} \times 0 = 0 \quad \text{--- (3)}$$

$$S_{13} \times S_{23}^* = 0$$

$$\boxed{S_{13} = 0} \Rightarrow \boxed{S_{23}^* = 0}$$

$$\text{from } \sum_{k=1}^N S_{ik} \cdot S_{jk}^* = 1$$

$$S_{12} \cdot S_{12}^* + S_{13} \cdot S_{13}^* = 1 \quad \text{--- (4)}$$

$$S_{12} \cdot S_{12}^* + S_{23} \cdot S_{23}^* = 1 \quad \text{--- (5)}$$

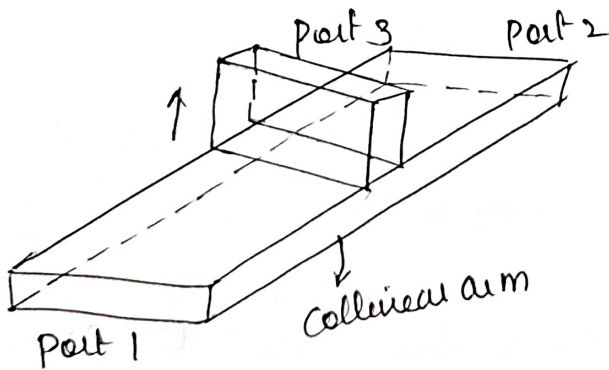
$$S_{13} \cdot S_{13}^* + S_{23} \cdot S_{23}^* = 1 \quad \text{--- (6)}$$

from above eq<sup>ns</sup> we can say that it is not possible to construct perfectly matched lossless reciprocal three port junction.

Q1.

(b)

S-matrix of E-plane Tee junction:



$$[S] = \begin{pmatrix} S_{11} & S_{12} & S_{13} \\ S_{21} & S_{22} & S_{23} \\ S_{31} & S_{32} & S_{33} \end{pmatrix}$$

1.  $S_{31} = S_{32}$

2.  $S_{33} = 0$  [perfectly matched port 3]

3.  $S_{23} = S_{32} = -S_{13} = -S_{31}$ ;  $S_{13} = S_{31}$ .

$$[S] = \begin{pmatrix} S_{11} & S_{12} & S_{13} \\ S_{12} & S_{22} & -S_{13} \\ S_{13} & -S_{13} & 0 \end{pmatrix}$$

$$[S][S^*] = [U]$$

$$\begin{bmatrix} S_{11} & S_{12} & S_{13} \\ S_{12} & S_{22} & -S_{13} \\ S_{13} & -S_{13} & 0 \end{bmatrix} \begin{bmatrix} S_{11}^* & S_{12}^* & S_{13}^* \\ S_{12}^* & S_{22}^* & -S_{13}^* \\ S_{13}^* & -S_{13}^* & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$S_{11} \cdot S_{11}^* + S_{12} \cdot S_{12}^* + S_{13} \cdot S_{13}^* = 1 \quad \text{--- (1)}$$

$$|S_{11}|^2 + |S_{12}|^2 + |S_{13}|^2 = 1$$

$$S_{12} \cdot S_{12}^* + S_{22} \cdot S_{22}^* + S_{13} \cdot S_{13}^* = 1$$

$$|S_{12}|^2 + |S_{22}|^2 + |S_{13}|^2 = 1 \quad \text{--- (2)}$$

$$S_{13} \cdot S_{13}^* + S_{13} \cdot S_{13}^* = 1 \quad \text{--- (3)}$$

$$|S_{13}|^2 + |S_{13}|^2 = 1$$

$$2|S_{13}|^2 = 1$$

$$|S_{13}| = \frac{1}{\sqrt{2}} \Rightarrow \boxed{S_{13} = \frac{1}{\sqrt{2}}}$$

from eq<sup>n</sup> ① & ②

$$|S_{11}| = |S_{22}| \Rightarrow \boxed{S_{11} = S_{22}}$$

$$2|S_{11}|^2 = 1 - \left(\frac{1}{\sqrt{2}}\right)^2$$
$$= 1 - \frac{1}{2}$$

$$2|S_{11}|^2 = \frac{1}{2} \Rightarrow \boxed{S_{11} = \frac{1}{2}}$$

$$[S] = \begin{pmatrix} S_{11} & S_{12} & S_{13} \\ S_{12} & S_{22} & -S_{13} \\ S_{13} & -S_{13} & 0 \end{pmatrix}$$

$$= \begin{pmatrix} \frac{1}{2} & \frac{1}{2} & \frac{1}{\sqrt{2}} \\ \frac{1}{2} & \frac{1}{2} & -\frac{1}{\sqrt{2}} \\ \frac{1}{\sqrt{2}} & -\frac{1}{\sqrt{2}} & 0 \end{pmatrix}$$

Q2.

S-parameter: It is a matrix which has  $n \times n$  matrix  
no of rows & no of columns are equal. The diagonal elements  
are 0, matched for perfectly matched condition.  
Symmetry property and Unitary property for loss  
less matrix.

Statement of Unitary property: The property states that the sum of terms in the matrix and its conjugate product is equal to 1

$$\sum_{k=1}^N S_{ki} S_{kj}^* = 1$$

Proof:

- Assume the impedance of all the ports are identical
- $Z_{0N} = 1$  (To be made)
- The current and voltage eq<sup>n</sup> is given by,

$$V_n = V_n^+ + V_n^- \quad \text{--- (1)}$$

$$I_n = I_n^+ - I_n^- \quad \text{--- (2)}$$

add eq<sup>n</sup> (1) & (2)

$$V_n + I_n = 2V_n^+$$

$$2[V^+] = [V] + [I]$$

$$[V^+] = \frac{1}{2} \{ [V] + [I] \}$$

$$[V^+] = \frac{1}{2} \{ [Z] + [U] \} [I] \quad \text{--- (3)}$$

Subtract eq<sup>n</sup> (2) & (1)

$$[V^-] = \frac{1}{2} \{ [Z] - [U] \} [I] \quad \text{--- (4)}$$

divide eq<sup>n</sup> (4) / (3)

$$\frac{[V^-]}{[V^+]} = \frac{\{ [Z] - [U] \} [I]}{\{ [Z] + [U] \} [I]}$$

$$[V^-] = [V^+] \{ [Z] - [U] \} \{ [Z] + [U] \}^{-1} \quad \text{--- (5)}$$

$$[S] = \{ [z] + [v] \} \{ [z] - [v] \}^{-1}$$

taking transpose on both side

$$[S]^t = \{ ([z] + [v])^t \} \{ ([z] - [v])^t \}^{-1}$$

Consider power

$$P_{av} = \frac{1}{2} \operatorname{Re} \{ [V^+]^t [I]^* \}$$

$$= \frac{1}{2} \operatorname{Re} \{ ([V^+]^t + [V^-]^t) ([V^+]^* - [V^-]^*) \}$$

$$= \frac{1}{2} \underbrace{[V^+]^t [V^+]^*}_{\downarrow} - \frac{1}{2} \underbrace{[V^-]^t [V^-]^*}_{\downarrow}$$

power produced  
but incident

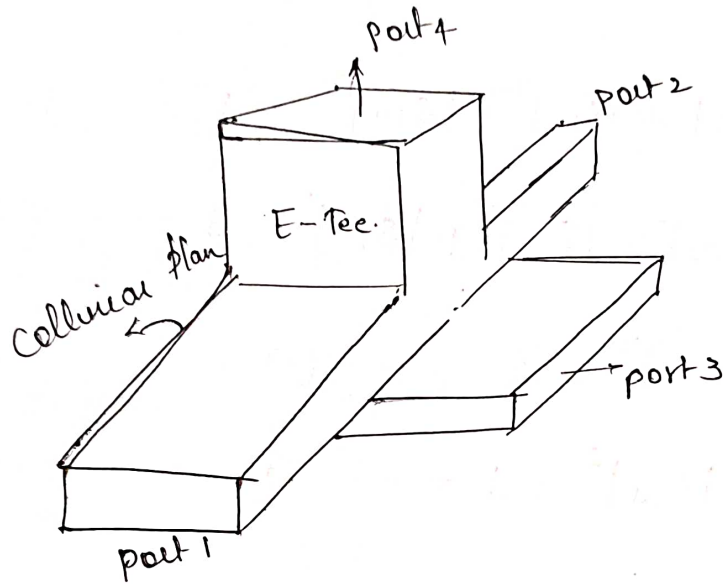
reflected power.

$$\sum_{k=1}^N S_{ki} S_{kj}^* = S_{ij}$$

where  $S_{ij} = \begin{cases} 1 & ; \quad i=j \\ 0 & ; \quad i \neq j \end{cases}$

$$\Rightarrow \underbrace{\sum_{k=1}^N S_{ki} S_{kj}^*}_{i \neq j} = 0 \quad \& \quad \underbrace{\sum_{k=1}^N S_{ki} S_{kj}^*}_{i=j} = 1$$

Q4. Magic Tee : It is a 4-port network combination of H Tee and E-Tee junction. They are symmetric, perfectly matched system. It is known as Magic Tee.



1.  $[S]$  matrix is  $4 \times 4 =$

$$\begin{pmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{21} & S_{22} & S_{23} & S_{24} \\ S_{31} & S_{32} & S_{33} & S_{34} \\ S_{41} & S_{42} & S_{43} & S_{44} \end{pmatrix}$$

2.  $S_{31} = S_{32}$  (H-Tee)

3.  $S_{41} = -S_{42}$  (E-Tee)

4.  $S_{33} = S_{44} = 0$  [perfectly matched cond<sup>n</sup>].

5. Symmetric  $S_{ij} = S_{ji}$

$$[S] = \begin{pmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{12} & S_{22} & S_{13} & -S_{14} \\ S_{13} & S_{13} & 0 & 0 \\ S_{14} & -S_{14} & 0 & 0 \end{pmatrix}$$



$$[S][S^*] = [U]$$

$$\begin{bmatrix} S_{11} & S_{12} & S_{13} & S_{14} \\ S_{12} & S_{22} & S_{13} & -S_{14} \\ S_{13} & S_{13} & 0 & 0 \\ S_{14} & -S_{14} & 0 & 0 \end{bmatrix} \begin{bmatrix} S_{11}^* & S_{12}^* & S_{13}^* & S_{14}^* \\ S_{12}^* & S_{22}^* & S_{13}^* & -S_{14}^* \\ S_{13}^* & S_{13}^* & 0 & 0 \\ S_{14}^* & -S_{14}^* & 0 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

$$|S_{11}|^2 + |S_{12}|^2 + |S_{13}|^2 + |S_{14}|^2 = 1 \quad \text{--- (1)}$$

$$|S_{12}|^2 + |S_{22}|^2 + |S_{13}|^2 + |S_{14}|^2 = 1 \quad \text{--- (2)}$$

$$|S_{13}|^2 + |S_{13}|^2 = 1 \quad \text{--- (3)}$$

$$|S_{14}|^2 + |S_{14}|^2 = 1 \quad \text{--- (4)}$$

from eq<sup>n</sup> (3)  $\Rightarrow$

$$2|S_{13}|^2 = 1$$

$$\boxed{S_{13} = \frac{1}{\sqrt{2}}}$$

from eq<sup>n</sup> (4)  $\Rightarrow$

$$2|S_{14}|^2 = 1$$

$$|S_{14}|^2 = \frac{1}{2}$$

$$\Rightarrow \boxed{S_{14} = \frac{1}{\sqrt{2}}}$$

from eq<sup>n</sup> ① & ②

$$|S_{11}|^2 = |S_{22}|^2 \Rightarrow \boxed{S_{11} = S_{22}}$$

Eq<sup>n</sup> ①  $\Rightarrow$

$$2|S_{11}|^2 + \left(\frac{1}{\sqrt{2}}\right)^2 + \left(\frac{1}{\sqrt{2}}\right)^2 = 1$$

$$\boxed{S_{11} = 0} \Rightarrow \boxed{S_{22} = 0}$$

$$\boxed{S_{12} = 0} \quad \&$$

$$\Rightarrow S_{\text{matrix}} = \begin{bmatrix} 0 & 0 & 1/\sqrt{2} & 1/\sqrt{2} \\ 0 & 0 & 1/\sqrt{2} & -1/\sqrt{2} \\ 1/\sqrt{2} & 1/\sqrt{2} & 0 & 0 \\ 1/\sqrt{2} & -1/\sqrt{2} & 0 & 0 \end{bmatrix}$$

Applications:

- $\rightarrow$  Used in Communication Laboratory
- $\rightarrow$  Digital Communication
- $\rightarrow$  To transmit the signal.

Q6. [S] & [Z] matrix relation.

$$\begin{aligned} \text{W.K.T. } V_1 &= Z_{11} I_1 + Z_{12} I_2 \quad \text{--- ①} \\ V_2 &= Z_{21} I_1 + Z_{22} I_2 \quad \text{--- ②} \end{aligned} \quad \left. \vphantom{\begin{aligned} V_1 &= Z_{11} I_1 + Z_{12} I_2 \\ V_2 &= Z_{21} I_1 + Z_{22} I_2 \end{aligned}} \right\} \text{Z-matrix}$$

$$\begin{aligned} \forall. V_1 &= A V_2 - B I_2 \quad \text{--- ③} \\ I_1 &= C V_2 - D I_2 \quad \text{--- ④} \end{aligned} \quad \left. \vphantom{\begin{aligned} V_1 &= A V_2 - B I_2 \\ I_1 &= C V_2 - D I_2 \end{aligned}} \right\} \text{ABCD matrix}$$

Rearranging eq<sup>n</sup> (2)

$$Z_{21} I_1 = V_2 - Z_{22} I_2$$

$$I_1 = \frac{1}{Z_{21}} V_2 - \frac{Z_{22}}{Z_{21}} I_2 \quad \text{--- (5)}$$

Compare eq<sup>n</sup> (5) & (4)

$$C = \frac{1}{Z_{21}} \quad \& \quad D = \frac{Z_{22}}{Z_{21}}$$

Replace (5) in (1)  $\Rightarrow$

$$V_1 = Z_{11} \left[ \frac{1}{Z_{21}} V_2 - \frac{Z_{22}}{Z_{21}} I_2 \right] + Z_{12} I_2$$

$$V_1 = \frac{Z_{11}}{Z_{21}} V_2 - \left( \frac{Z_{11} Z_{22}}{Z_{21}} + Z_{12} \right) I_2 \quad \text{--- (6)}$$

Compare eq<sup>n</sup> (6) with (3)

$$A = \frac{Z_{11}}{Z_{21}} \quad \& \quad B = \frac{Z_{11} Z_{22}}{Z_{21}} + Z_{12}$$

$$I_1 = Y_{11} V_1 + Y_{12} V_2 \quad \text{--- (7)}$$

$$I_2 = Y_{21} V_1 + Y_{22} V_2 \quad \text{--- (8)}$$

Q7.

[S] and [ABCD] matrix

$$V_1 = Z_{11} I_1 + Z_{12} I_2 \quad \text{--- (1)}$$

$$V_2 = Z_{21} I_1 + Z_{22} I_2 \quad \text{--- (2)}$$

$$V_1 = AV_2 - BI_2 \quad \text{--- (3)}$$

$$I_1 = CV_2 - DI_2 \quad \text{--- (4)}$$

rearrange eq<sup>n</sup> (2)  $\Rightarrow$

$$V_2 - Z_{22} I_2 = Z_{21} I_1$$

$$I_1 = \frac{V_2}{Z_{21}} - \frac{Z_{22}}{Z_{21}} I_2 \quad \text{--- (5)}$$

$$\Rightarrow C = \frac{1}{Z_{21}} \quad \& \quad D = \frac{Z_{22}}{Z_{21}}$$

replace (5) in eq<sup>n</sup> (1)

$$V_1 = Z_{11} \left[ \frac{1}{Z_{21}} V_2 - \frac{Z_{22}}{Z_{21}} I_2 \right] + Z_{12} I_2$$

$$V_1 = \frac{Z_{11}}{Z_{21}} V_2 - \left( \frac{Z_{11} Z_{22}}{Z_{21}} + Z_{12} \right) I_2 \quad \text{--- (6)}$$

Compare (6) & (3)

$$A = \frac{Z_{11}}{Z_{21}} \quad \& \quad B = \frac{Z_{11} Z_{22}}{Z_{21}} + Z_{12}$$

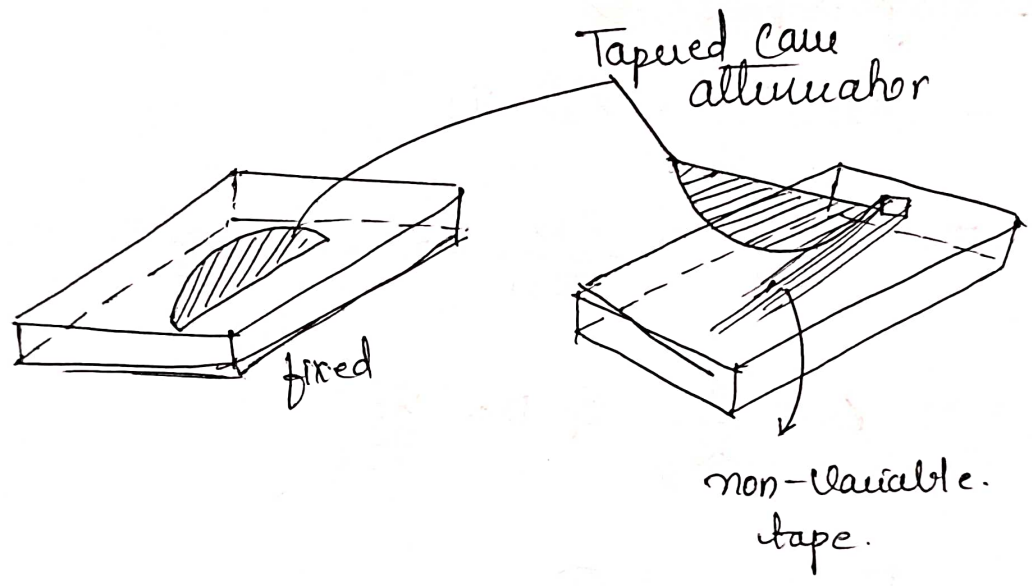
Q5.

Attenuator : work as dissipator / variable.

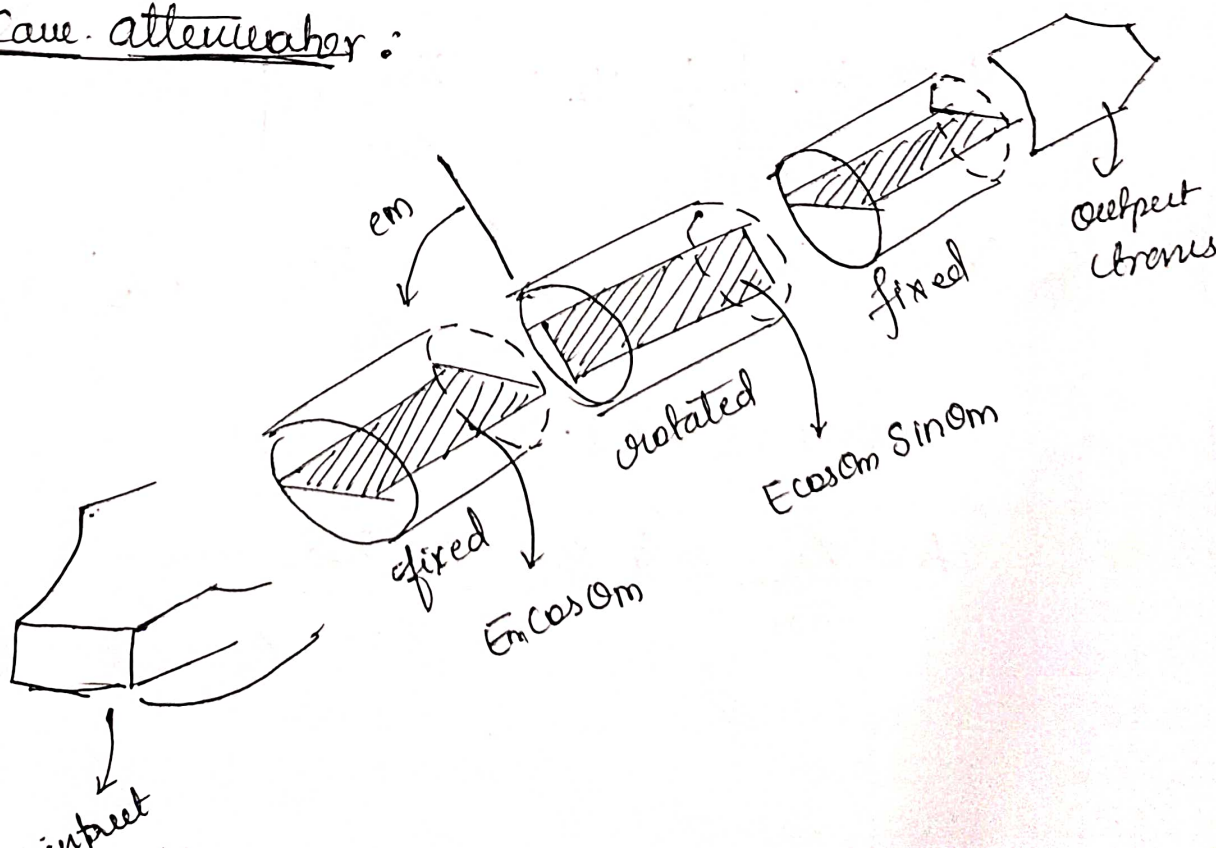
There are two types of attenuators:

- Resistive <sup>caus</sup> attenuator.
- Rotate caus attenuator

Resistive caus attenuator : It can be fixed / variable.



Rotate Caus attenuator :



- There are three main successive parts two fixed and one rotated
- There are two ~~see~~ trans. input trans and the output trans where there will less SWR.
- Attenuation can be controlled by rotating.
- The mini attenuation can be obtained with  $\theta_m = 0^\circ$ .
- The max attenuation can be obtained when  $\theta_m = 90^\circ$ .

part	attenuation value (dB)
fixed	40
Rotated	80.

→ Attenuation can be obtained by formula

$$A_T = 20 \log \left( \frac{1}{\cos^4 \theta_m} \right)$$