CMR INSTITUTE OF **TECHNOLOGY**

Internal Assessment Test - II

P.T.O

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1.
$$
x_1(n) = (3, 2, 1, 2)
$$

\n $x_1(n) = (3, 2, 1, 2)$
\n $x_2(n) = (1, 2, 1, 2)^{(1, 0)(1, 0) + (1, 1, 0, 0, 0)}$
\n $x_2(n) = (1, 2, 1, 2)^{(1, 0)(1, 0) + (1, 1, 0, 0, 0)}$
\n $x_1 \rightarrow 0$ to x_3
\n $x_2 \rightarrow 0$ to x_3
\n $x_3 \rightarrow 0$ to x_3
\n $x_4 \rightarrow 0$ to x_3
\n $x_5 \rightarrow 0$ to x_3
\n $x_6 \rightarrow 0$
\n $x_7 \rightarrow 0$ to x_3
\n $x_8 \rightarrow 0$ to x_3
\n $x_9 \rightarrow 0$
\n $x_1 \rightarrow 0$ to x_3
\n $x_1 \rightarrow 0$ to x_1
\n $x_2 \rightarrow 0$ to x_3
\n $x_3 \rightarrow 0$ to x_3
\n $x_4 \rightarrow 0$ to x_3
\n $x_5 \rightarrow 0$ to x_3
\n $x_6 \rightarrow 0$ to x_1
\n $x_7 \rightarrow 0$ to x_8
\n $x_9 \rightarrow 0$ to x_9
\n $x_1 \rightarrow 0$ to $x_1 \rightarrow 0$
\n $x_1 \rightarrow 0$ to $x_1 \rightarrow 0$
\n $x_1 \rightarrow 0$ to $x_1 \$

 $2.$

Ю

$$
h(t) = t u(t)
$$

We know that the step response is given by.

$$
s(t) = \int_{-\infty}^{t} h(\tau) d\tau
$$

=
$$
\int_{0}^{t} \tau^{2} d\tau
$$

=
$$
\frac{\tau^{3}}{2} \bigg|_{0}^{t}
$$

=
$$
\frac{\tau^{3}}{2} \qquad \forall t \ge 0
$$

∴ The step response s(t) = $\frac{t^3}{2}u(t)$

$$
s(n) = \sum_{k=-\infty}^{n} h(k)
$$

$$
For n < 0 \quad ; s(n) = 0
$$

For
$$
n \ge 0
$$
 : $s(n) = \sum_{k=0}^{\infty} (\frac{1}{2})^k$
\n
$$
s(n) = \frac{1 - (\frac{1}{2})^{n+1}}{1 - \frac{1}{2}}
$$
\n
$$
= 2[1 - (\frac{1}{2})^{n+1}] \qquad \text{in } 0
$$
\n
$$
s(n) = 2[1 - (\frac{1}{2})^{n+1}] \cdot u(n)
$$

$r^2 + 6r + 8 = 0$ $x^2 + 4y + 2y + 8 = 0$ $y^{k}(t) = c_1 e^{4t} + c_2 e^{2t}$ $y(x) = 2$
 $y(x) = 2$
 $x = c_1 + c_2$ $\frac{dy(t)}{dt} = -4c_1 e^{4t} - 2l_2 e^{2t}$ x^2
3 = $-4C_1 - 2C_2$ $C_{1} = -\frac{a_{1}}{2}$ $C_{2} = 1$ $4xC_{1}$ $y^{n}(t) = \frac{1}{a}e^{4t} + \frac{1}{a}e^{2t}$

 $7ub = \frac{1}{a}e^{-\frac{1}{a}t}$
 $7ub = \frac{1}{e}tut$ $y''t = K\frac{1}{e}t$
 $\frac{1}{e}k\frac{1}{e} + b\frac{1}{e}k\frac{1}{e} + 8k\frac{1}{e}t = \frac{1}{e} - a\frac{1}{e}t$
 $+ke^{-t} - bk\frac{1}{e}t + 8k\frac{1}{e}t = -e^{-a}e^{-b}$
 $+ke^{-t} - bk\frac{1}{e}t = -3e^{-b}$
 $1e^{a} + b = -4e^{-b}$
 $1e^{a} + b = -4$ $\frac{1}{dt}(\overline{e}^{\overline{t}})+\lambda\overline{e}^{\overline{t}}$ $y^{f}(t) = 5e^{-4t} + 4e^{-2t} + e^{2t}$

3.

$$
y(a) = 0 \quad \frac{dy(d)}{dt} = 0
$$
\n
$$
0 = C_{3} + C_{4} + 1
$$
\n
$$
\frac{dy(d)}{dt} = -4C_{3}e^{4t} - 1e^{4t} - e^{4t} - e^{4t}
$$
\n
$$
0 = -4C_{3} - 1e^{4t} - 1
$$
\n
$$
C_{3} + C_{4} = -1
$$
\n
$$
C_{4} - 1e^{4t} - 1
$$
\n
$$
C_{5} - 1e^{4t} - 1
$$
\n
$$
C_{6} - 1e^{4t} - 1
$$
\n
$$
C_{7} - 1e^{4t} - 1e^{4t}
$$
\n
$$
y(t) = \int_{0}^{t} 1e^{4t} - 1e^{4t} - 1e^{4t} - 1e^{4t} - 1e^{4t} - 1e^{4t}
$$
\n
$$
-1e^{4t} + 1e^{4t} + 1e^{4t} - 1e^{4t} - 1e^{4t} + 1e^{4t}
$$
\n
$$
-3e^{4t} + 1e^{4t} + 1e^{4t} + 1e^{4t} + 1e^{4t}
$$
\n
$$
y(t) = -3e^{4t} + 1e^{4t} + 1e^{4t} + 1e^{4t}
$$

4.

5.

2.3.2 The Distributive Property

Another basic property of convolution is 'distributive property'.

i.e. In discrete-time $x(n) * {h₁(n) + h₂(n)} = x(n) * h₁(n) + x(n) * h₂(n)$ and in continuous-time,

 $x(t) * {h_1(t) + h_2(t)} = x(t) * h_1(t) + x(t) * h_2(t)$

Consider an interconnection of discrete-time LTI system as shown in Fig. 2.8

2.3.3 The Associative Property

This is another important property of convolution i.e. in discrete-time,
 $x(n) * \{h_1(n) * h_2(n)\} = \{x(n) * h_1(n)\} * h_2(n)$ and in continuous-time, $x(t) * {h_1(t) * h_2(t)} = {x(t) * h_1(t)} * h_2(t)$

Fig. 2.13 below.

$$
x(t) \longrightarrow \boxed{h_i(t) * h_2(t)} \longrightarrow y(t)
$$

Fig. 2.13

Hence the impulse response of two LTI system connected in cascade is the convolution of the individual impulse responses.

Also we know that convolution is commutative. Therefore we can write the system in Fig. 2.13 as below.

$$
x(t) \longrightarrow \boxed{h_2(t) * h_1(t)} \longrightarrow y(t)
$$

Fig. 2.14

$$
x(t) \longrightarrow \boxed{h_2(t)} \longrightarrow \boxed{h_1(t)} \longrightarrow y(t)
$$

From the above discussion we conclude that the output of a cascade combination of LTI systems is independent of the order in which the systems are connected.

Similarly, for discrete-time we have,

$$
x(n) * {h1(n) * h2(n)} = x(n) * {h2(n) * h1(n)}
$$

2.3.5 Causal System

We know that the output of a causal system depends only on the present and/or past values of the input to the system. i.e. for a discrete-time LTI system to be causal, its $h(n) = 0$ for $n < 0$. Similarly, for a continuous-time LTI system to be causal, its $h(t) = 0$ for $t < 0$.

6.

Here
$$
x(\tau) = u(\tau) - u(\tau - 3)
$$

\n
$$
= \begin{cases} 1 & \text{for } 0 \le \tau \le 3 \\ 0 & \text{elsewhere} \end{cases}
$$
\nand
$$
h(\tau) = u(\tau) - u(\tau - 2)
$$
\n
$$
= \begin{cases} 1 & \text{for } 0 \le \tau \le 2 \\ 0 & \text{elsewhere} \end{cases}
$$

Fig. 2.2.34 shows $x(t)$ and $h(t)$. It also shows $h(t-\tau)$ for various values of t. Output is given as $y(t) = \int x(\tau) \cdot h(t-\tau) d\tau$. The overlap of $x(\tau)$ and $h(t-\tau)$ is shown for various values of t.

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Case-I: $t \le 0$ and $t \ge 5$

For this case there is no overlap betwee $x(\tau)$ and $h(t-\tau)$. Hence $y(t) = 0$.

Case-II: $0 \le t \le 2$ Refer Fig. (d)

$$
y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau = \int_{0}^{t} 1 d\tau = t
$$

Case-III : $2 \le t \le 3$ Refer Fig. (e)

$$
y(t) = \int_{t-2}^{t} 1 dt = 2
$$

Case-IV : $3 \le t \le 5$ Refer Fig. (f)

$$
y(t) = \int_{t-2}^{3} 1 dt = 5 - t
$$

Thus the result will be,

$$
y(t) = \begin{cases} t & \text{for } 0 \le t \le 2 \\ 2 & \text{for } 2 \le t \le 3 \\ 5-t & \text{for } 3 \le t \le 5 \end{cases}
$$