

CO2:	Explain the need of operating transformers in parallel and the procedure to do it.	3	3	2	2	-	-	-	-	-	-	-	-
CO3:	Illustrate the concept of auto transformer; tap changing transformer and tertiary winding.	3	2	2	1	-	-	-	-	-	-	-	-
CO4:	Analyze armature reaction and commutation and their effects in a dc machine.	3	3	2	1	-	-	-	-	-	-	-	-
CO5:	Describe the construction, operation, characteristics and applications of synchronous generators	3	2	2	2	-	-	-	-	-	-	-	-
CO6:	Perform the analysis of synchronous machines by using different methods	3	3	2	2	-	-	-	-	-	-	-	-

Cognitive level	KEYWORDS
L1	List, define, tell, describe, identify, show, label, collect, examine, tabulate, quote, name, who, when, where, etc.
L2	summarize, describe, interpret, contrast, predict, associate, distinguish, estimate, differentiate, discuss, extend
L3	Apply, demonstrate, calculate, complete, illustrate, show, solve, examine, modify, relate, change, classify, experiment, discover.
L4	Analyze, separate, order, explain, connect, classify, arrange, divide, compare, select, explain, infer.
L5	Assess, decide, rank, grade, test, measure, recommend, convince, select, judge, explain, discriminate, support, conclude, compare, summarize.

PO1 - *Engineering knowledge*; PO2 - *Problem analysis*; PO3 - *Design/development of solutions*; PO4 - *Conduct investigations of complex problems*; PO5 - *Modern tool usage*; PO6 - *The Engineer and society*; PO7- *Environment and sustainability*; PO8 - *Ethics*; PO9 - *Individual and team work*; PO10 - *Communication*; PO11 - *Project management and finance*; PO12 - *Life-long learning*

SOLUTION

1. a.)

Conditions for Parallel Operation

When two or more transformers are to be operated in parallel, then certain conditions have to be met for proper operation. These conditions are -

- **Voltage ratio of all connected transformers must be same.**
If the voltage ratio is not same, then the secondaries will not show equal voltage even if the primaries are connected to same busbar. This results in a circulating current in secondaries, and hence there will be reflected circulating current on the primary side also. In this case, considerable amount of current is drawn by the transformers even without load.
- **The per unit (pu) impedance of each transformer on its own base must be same.**
Sometimes, transformers of different ratings may be required to operate in parallel. For, proper load sharing, voltage drop across each machine must be same. That is, larger transformer has to draw equivalent large current. That is why per unit impedance of the connected transformers must be same.
- **The polarity of all connected transformers must be same in order to avoid circulating currents in transformers.** Polarity of a transformer means the instantaneous direction of induced emf in secondary. If polarity is opposite to each other, huge circulating current flows.
- **The phase sequence must be identical of all parallel transformers.**
This condition is relevant to poly-phase transformers only. If the phase sequences are not same, then transformers cannot be connected in parallel.
- **The ratio of their winding resistances to reactances should be equal for both the transformers.**
This condition ensures that both transformers operate at the same power factor, thus sharing their active power and reactive volt-amperes according to their ratings.

Parallel Operation of 3-phase Transformers

All the conditions which apply to the parallel operation of single-phase transformers also apply to the parallel running of 3-phase transformers but with the following additions :

1. The voltage ratio must refer to the terminal *voltage of primary and secondary*. It is obvious that this ratio may not be equal to the ratio of the number of turns per phase.
2. The phase displacement between primary and secondary voltages must be the same for all transformers which are to be connected for parallel operation.
3. The phase sequence must be the same.
4. All three transformers in the 3-phase transformer bank will be of the same construction either core or shell.

b)

1b) $E_A = 6600V$ $E_B = 6400V$
 $Z_A = 0.3 + j3$ $Z_B = 0.2 + j2$ $Z_L = 8 + j6$

$$I_A = \frac{E_A Z_B + (E_A - E_B) Z_L}{Z_A Z_B + (Z_A + Z_B) Z_L} = \frac{6600(0.2 + j2) - (6600 - 6400)(8 + j6)}{(0.2 + j2)(0.3 + j3) + (0.2 + j2 + 0.3 + j3)(8 + j6)}$$

$$I_B = \frac{E_B Z_A - (E_A - E_B) Z_L}{Z_A Z_B + (Z_A + Z_B) Z_L} = \frac{6400(0.3 + j3) - (6600 - 6400)(8 + j6)}{(0.2 + j2)(0.3 + j3) + (0.2 + j2 + 0.3 + j3)(8 + j6)}$$

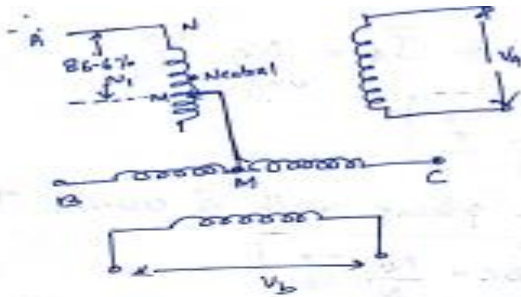
$$I_A = 267.44 \angle -27.31^\circ \text{ A}$$

$$= 182.67 - 198.06j \text{ A}$$

$$I_B = 264.09922 - 198.08437j \text{ A}$$

$$= 330.13 \angle -36.87^\circ \text{ A}$$

2.

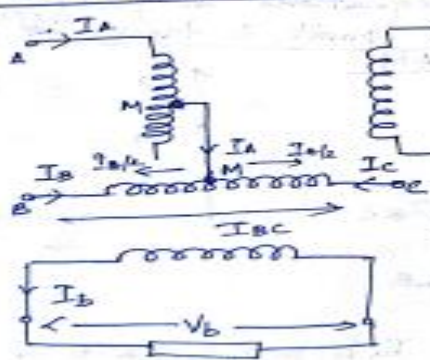


So to make $|V_A| = |V_B|$

- Teaser transformer turns ratio must be $\sqrt{3}/2 N_1 / N_2$
- Main transformer turns ratio is N_1 / N_2

Load Conditions

Balanced load. (Resistive load)



when a balanced resistive load is connected across the secondaries,

- I_A, I_B is the teaser and main transformer 2^o current
- I_A, I_B, I_C is the 1^o currents input to the terminals A, B, C for teaser and main transformer.

• I_{BC} is the total 1^o current in main transformer

Then from the fig:-

$$\vec{I}_A - \frac{N_2}{N_1} \vec{I}_a \Rightarrow \vec{I}_A = \frac{N_2}{\sqrt{3}/2 N_1} \vec{I}_a$$

$$\text{or } \left[\vec{I}_A = \frac{2}{\sqrt{3}} \frac{N_2}{N_1} \vec{I}_a \right]$$

ie: \vec{I}_A is in phase with 2nd current \vec{I}_a .

$$\text{Similarly, } \left[\vec{I}_{BC} = \frac{N_2}{N_1} \vec{I}_b \right]$$

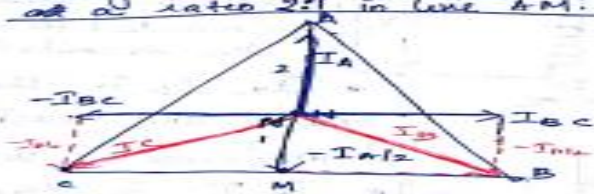
ie: \vec{I}_{BC} is in phase with 2nd current \vec{I}_b .

$$\text{Also } \vec{I}_B = \vec{I}_{BC} - \vec{I}_A/2$$

[As M is the midpoint I_A divides equally at point M]

$$\vec{I}_C = -\vec{I}_{BC} - \vec{I}_A/2$$

To represent these equations in phasor diagram, the Neutral point is taken at point with a ratio 2:1 in line AM.



- As load is resistive I_a is in phase with V_a
 I_b is in phase with V_b
 Draw I_A in direction I_a & I_{BC} in direction I_b

Balanced inductive load

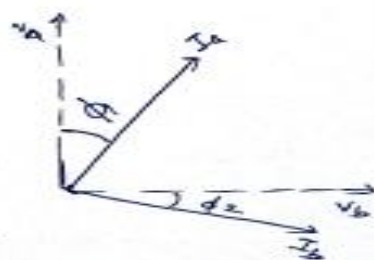
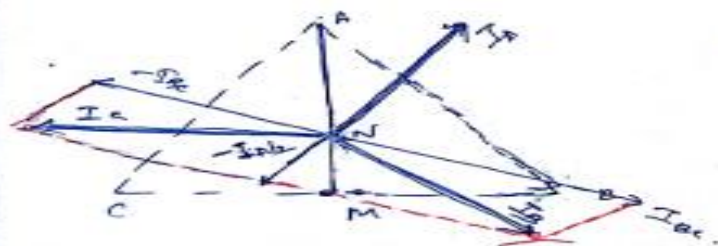
For balanced inductive load

I_a lags and I_b lags V_a and V_b by some angle ϕ corresponding to load. Then the phasor diagram will be as follows.



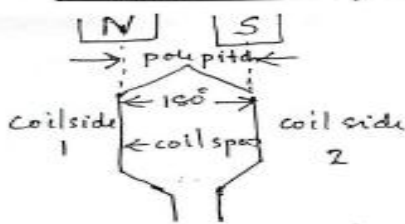
Unbalanced Load

For unbalanced load. I_a lags V_a by ϕ_1 angle ϕ_1 and I_b lag V_b by angle ϕ_2 and $\phi_1 \neq \phi_2$.



3. a.

Full pitched wdg



Here coil span = pole pitch = 180°

E_{c1} → emf in coil side 1
 E_{c2} → emf in coil side 2
 E_{c1} & E_{c2} are in phase with each other for full pitched winding.



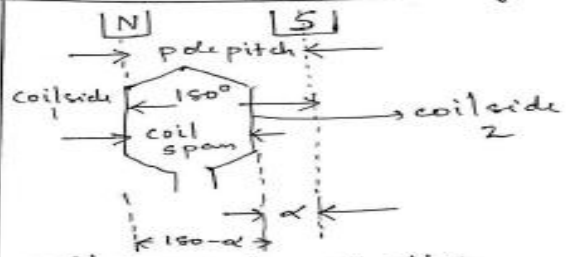
Resultant emf



if $|E_{c1}| = |E_{c2}| = E$

$$E_c = 2E$$

Short pitched wdg.

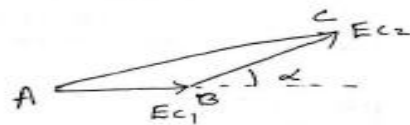


coil span < pole pitch
 coil span $\neq 180^\circ$

α → short pitched angle

coil span = $180 - \alpha$

E_{c1} & E_{c2} are not in phase.
 E_{c2} is displaced by ' α '



Resultant emf

is AC

if short pitched coils are used, there exists coil span factor or Pitch factor K_c .

$$K_c = \frac{\text{emf induced with short pitched wdg.}}{\text{emf induced with full pitched wdg.}}$$

$$= \frac{\text{Phasor sum of emf in coils}}{\text{Arithmetic sum of emf in coils}}$$

$$= \frac{AC}{2E}$$

To find AC



$$\therefore AD = AB \cos \alpha/2$$

$$AD = E_{c1} \cos \alpha/2$$

$$AC = AD + DC = 2AD = 2E \cos \alpha/2$$

$$\left. \begin{array}{l} \text{Coil span factor} \\ \text{Pitch factor} \end{array} \right\} = K_c = \frac{AC}{2E} = \frac{2E \cos \alpha/2}{2E}$$

$K_c = \cos \alpha/2$ where α is the short pitch angle.

From ΔADB

$\angle DAB = \alpha/2$
 (External angle is sum of internal angles)

$$|E_{c1}| = |E_{c2}| = E$$

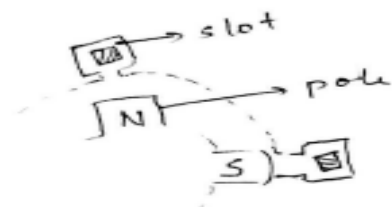
Concentrated and Distributed winding.

a) concentrated winding

In this type,

No. of Poles = No. of slots
= no. of coilsides

i.e; slots / pole = 1

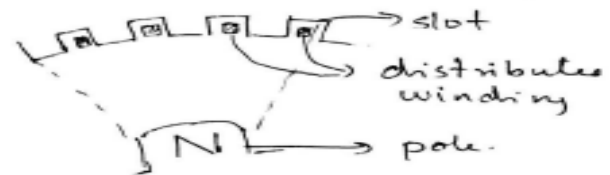


One coil side is inside one slot under one pole and the other coil side is inside other slot under next pole. emf induced in all coilsides (E_{c1}, E_{c2}, \dots) are in phase.

b) Distributed winding

In this type

No. of slots $>$ no. of poles.
Conductors are placed in several slots under one pole



Advantages of distributed wd

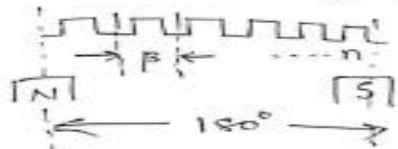
1. Gives sinusoidal emf
2. Reduces harmonics & noise
3. Reduces armature reaction
4. Cooling is effective and easy.

emf induced in one (E_{c1}) coil side is shifted from other by β (slot angle)

slot angle (β)

Let no. of slots / pole = n

$$\text{slot angle } \beta = \frac{180^\circ}{n} = \frac{180^\circ}{\text{slots/pole}}$$



Electrical degree corresponds to one slot is the slot angle.

Distribution factor or Breadth factor K_d

Let no of slots / pole / phase = m

$$\text{slot angle } \beta = \frac{180^\circ}{\text{slots/pole}}$$

Let AB, BC, CD, EF are the coilsides in each slot, (total No = m)

For distributed winding, emf induced in all slots has a phase difference of β .



Length $OA = r$
(radius of the sector)
 N is the midpoint of AB

Consider ΔOAN $\angle AON = \beta/2$

$$AN = r \sin \beta/2$$

$$AB = 2AN = 2r \sin \beta/2$$

For m no. of slots
 Algebraic sum of ^{coil} emfs = $2m\alpha \sin \beta/2$ ———— (1)

Draw the resultant vector AF.

Let 'P' is the mid point of AF

Then

$$\angle AOP = \frac{m\beta}{2}$$

In ΔAOP

$$AP = OA \sin \left(\frac{m\beta}{2} \right) = \alpha \sin \left(\frac{m\beta}{2} \right)$$

$$\therefore AF = 2AP = 2\alpha \sin \frac{m\beta}{2}$$

\therefore Distribution factor

$$K_d = \frac{\text{Vector sum of coil emfs}}{\text{Algebraic sum of coil emfs}}$$

$$= \frac{2\alpha \sin \left(\frac{m\beta}{2} \right)}{2m\alpha \sin \left(\beta/2 \right)}$$

$$= \frac{\sin \left(\frac{m\beta}{2} \right)}{m \sin \left(\beta/2 \right)}$$

b)

$$3b) E_{ph} = 2.22 f \Phi Z_{ph} K_c K_d \text{ Volts.}$$

$$f = \frac{1200 \cdot PN}{120} = \frac{8 \times 750}{120} = 50 \text{ Hz}$$

$$\Phi = 0.06 \text{ Wb}$$

$$Z_{ph} = \frac{72 \times 12}{3} = \underline{288}$$

$$\text{pole pitch} = \text{no. of slots/pole} = 72/9 = 8$$

$$\alpha = \frac{2}{9} \times 180 = 40$$

$$K_c = \cos(\alpha/2) = \cos 20 = 0.9397$$

$$m = \text{no. of slots/pole/phase} = 72/9/3 = 2.7 \approx 2.7$$

$$\beta = \frac{180}{\text{pole pitch}} = \frac{180}{8} = \underline{22.5}$$

$$K_d = \frac{\sin \left(\frac{2.7 \times 22.5}{2} \right)}{2.7 \sin \left(\frac{22.5}{2} \right)} = \frac{0.5056}{0.8267} = 0.6117$$

$$E_{ph} = 1729.96 \text{ V}$$

$$E_L = 2996.38 \text{ V}$$

$$1710.85 \text{ V}$$

$$2963.27 \text{ V}$$

4. a)

Advantages of chorded pitch/short pitched winding

- It saves copper of end connections
- Reduction in resistance and inductance of the winding due to the lesser length of the coil ends
- The wave form of the induced emf is improved
- The distorting harmonics can be reduced
- Due to elimination of high frequency harmonics eddy current and hysteresis losses are reduced, thereby increasing the efficiency
- Mechanical strength of coil is increased

Advantages of distribution Winding

- Harmonics are reduced
- Induced voltage approached sinusoidal wave form
- Armature reaction effect is reduced
- Losses are reduced
- Efficiency is improved
- Provide better cooling

At leading pf load

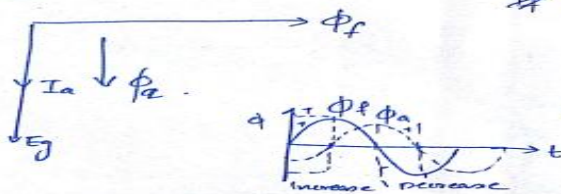


5.

* ARMATURE REACTION IN ALTERNATORS

when load is connected to the alternator terminal a current flows in the armature winding which produces armature flux. This flux interacts with the field flux and give different effects on the induced emf. This effect of armature flux on main flux is known as armature reaction. Armature reaction varies for loads at different power factors.

(i). At upf load (a Resistive load)

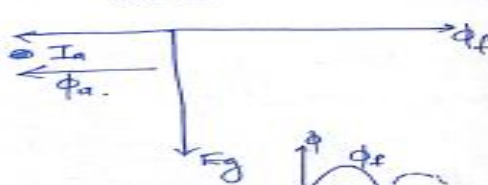


* Let ϕ_f = main field flux
 E_g = Generated emf
 I_a = Armature current
 ϕ_a = Armature flux.

- * The generated emf E_g lags the main field flux by 90°
- * As the connected load is resistive E_g and I_a are in phase.
- * Armature flux ϕ_a will therefore be in phase with I_a and lags ϕ_f by 90° .

So the resultant flux $\phi_r = \sqrt{\phi_f^2 + \phi_a^2}$ will vary non-uniformly, i.e; at some portions ϕ_r will increase and at some other times it will decrease. i.e., resultant flux has a distorted nature. This effect on armature flux on main flux is called cross magnetizing effect.

At lagging pf load (a inductively load)

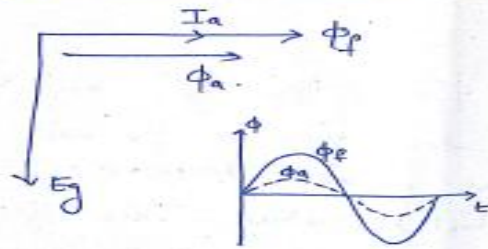


* when inductively load is connected armature current will lag voltage by 90°
~~so not phi~~
 * So armature flux ϕ_a is in phase with I_a will lag ϕ_f by 180°

* $\therefore \phi_a$ will be direct phase opposition with ϕ_f . So resultant flux will be the difference of these two fluxes.

* Net flux ϕ_r has lower magnitude. So this effect is called 'demagnetizing effect'.

iii) Leading pf load (Pure Capacitive load)



* For capacitive load, the armature current will lead the voltage by 90° .

* So I_a will be in phase with ϕ_f and so is ϕ_a .

* $\therefore \phi_f$ and ϕ_a are in phase & they have additive nature and net flux ϕ_r has higher magnitude. So this effect is called 'Magnetizing effect'.

In practical cases as loads are partially resistive, inductive or capacitive in nature, that these effects of armature reaction varies. Usually net effect of armature reaction is considered as a voltage drop. So to represent this, the armature reaction is represented as a reactance called armature reactance ~~drop~~ (X_a) and reduction of voltage is represented as voltage drop across it ($I_a X_a$)

6. ,

Voltage Regulation in Alternator

Voltage regulation is the change of alternator voltage from no load to full load.

$$R = E_0 - V$$

Regulation is always represented as percentage.

$$\therefore \% R = \frac{E_0 - V}{V} \times 100$$

The value of regulation depends on the load power factor.

* For resistive load,

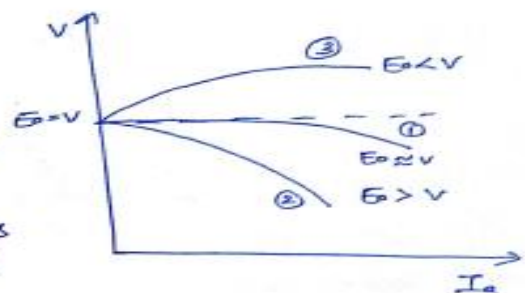
$$E_0 \approx V \text{ So } \% R = 0 \text{ or } +ve \text{ (low value)}$$

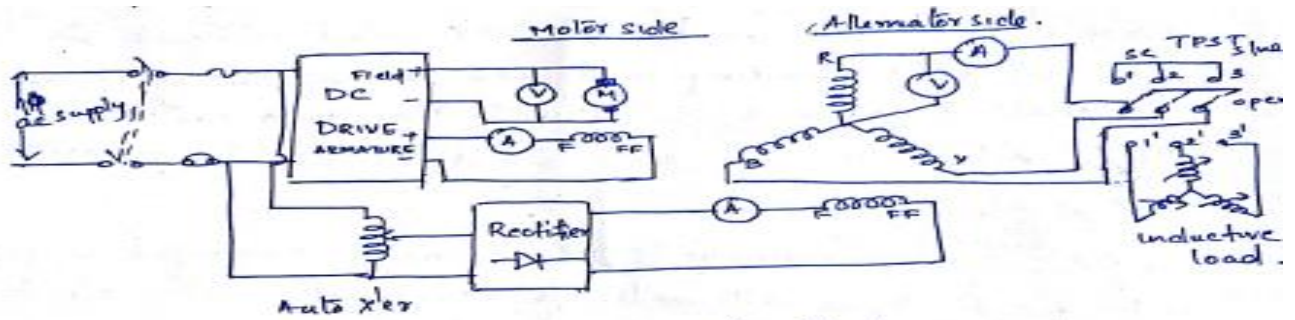
* For inductive load

$$E_0 > V \text{ So } \% R = +ve \text{ always (Demagnetizing)}$$

* For capacitive load,

$$E_0 < V \text{ So } \% R = -ve \text{ always (Magnetizing)}$$



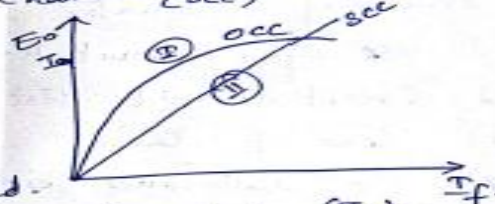


- * Connections are made as per circuit diagram.
- * TPST switch is in open or no-contact position.
- * Turn on the supply and set rated speed to the dc motor by adjusting dc drive.
- * OC test

Vary the alternator field supply and note different voltage values across alternator armature with TPST in open condition.

Tabulate the no load voltages and corresponding field current. The test can be conducted up to 125% of rated voltage. A graph can be plotted with the data which is known as open circuit characteristics as shown as (I).

SC test
With field minimum in alternator, close the TPST to position 1-2-3, the armature winding is now short circuited.



Now adjust the field and set rated current in the armature winding and note corresponding field current. A graph is plotted with this data as shown in (II) which is known as short circuit characteristics (SCC).

(i) Synchronous impedance method

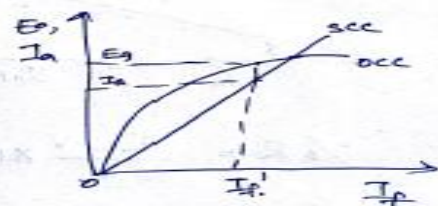
This method is also known as emf method. Here the magnetic circuit is assumed to be unsaturated. In this method MMFs (fluxes) produced by rotor and stator are replaced by their equivalent emf and hence called emf method. Regulation calculated by this method has high value compared to actual value of regulation and so this method is also called pessimistic method.

To calculate regulation oc test, sc test and dc resistance test of alternator is conducted. From the OCC and SCC

$$Z_s = \frac{\text{OC voltage}}{\text{SC current}} \quad \text{At same } I_f$$

$$R_a = 1.6 \times R_{dc}$$

$$\therefore \text{Synchronous reactance } X_s = \sqrt{Z_s^2 - R_a^2}$$



Regulation is to be calculated for different loads

Case 1 Lagging pf load (inductive load)

Let V_{ph} = Terminal voltage / phase

I_a = Armature current

ϕ = angle ~~between~~ current lags with voltage

The phasor diagram is shown.

To calculate E_o from AOC

Such that

$$OC^2 = OD^2 + DC^2$$

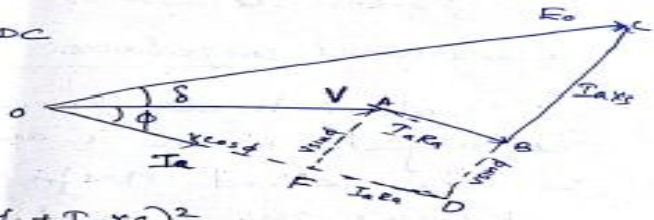
$$= (OF + FO)^2 + (DB + BC)^2$$

$$= (V \cos \phi + I_a R_a)^2 + (V \sin \phi + I_a X_s)^2$$

$$\therefore E_o = \sqrt{(V \cos \phi + I_a R_a)^2 + (V \sin \phi + I_a X_s)^2} \quad \text{--- (1)}$$

$$\therefore \%R = \frac{E_o - V}{V} \times 100$$

where E_o = no load voltage / phase



Case 2 :- Upf load (resistive load)



For the load $\phi = 0$

$$\therefore \cos \phi = 1 \quad \sin \phi = 0$$

From eqn (1)

$$E_o = \sqrt{(V + I_a R_a)^2 + (I_a X_s)^2}$$

$$\%R = \frac{E_o - V}{V} \times 100$$

Case 3 Leading pf load (capacitive load)

For leading pf load $X_s = -ve$

$$\therefore E_o = \sqrt{(V \cos \phi + I_a R_a)^2 + (V \sin \phi - I_a X_s)^2}$$

$$\%R = \frac{E_o - V}{V} \times 100$$



7.

7) $R_a = 0.2 \Omega$ $V = \frac{2300}{\sqrt{3}} = 1327.9 V$

$X_s = \frac{780/\sqrt{3}}{150} = 3.00 \Omega$ $I_a = 25 A (I_L)$

$X_s = \sqrt{3.00^2 + 0.2^2} = 3.2$

At lagging pf 0.8

$$E_o = \sqrt{(V \cos \phi + I_a R_a)^2 + (V \sin \phi + I_a X_s)^2}$$

$$= \sqrt{(1327.9 \times 0.8 + 25 \times 0.2)^2 + (1327.9 \times 0.6 + 25 \times 3)^2}$$

$$= 1378.09 V \quad \%R = 3.78 \%$$

At leading pf (0.6)

$$E_o = \sqrt{(V \cos \phi + I_a R_a)^2 + (V \sin \phi - I_a X_s)^2}$$

$$= \sqrt{(1327.9 \times 0.6 + 25 \times 0.2)^2 + (1327.9 \times 0.8 - 25 \times 3)^2}$$

$$= 1271.28 V$$

$$\%R = -4.26 \%$$

At upf

$$E_o = \sqrt{(V + I_a R_a)^2 + (I_a X_s)^2}$$

$$= \sqrt{(1327.9 + 25 \times 0.2)^2 + (25 \times 3)^2}$$

$$= 1335 V$$

$$\%R = 0.535 \%$$